

Design of FIR Filters

- Unlike IIR filter design, FIR filter design is directly performed in the discrete domain.
- Two approaches:

Window method and Minimax method

- Window method:

Given a desired frequency response $H_d(e^{j\omega})$;

Compute the desired impulse response $h_d(n)$;

Multiply the desired impulse response by a window to yield a finite impulse response $h(n) = h_d(n)w(n)$;

- To ensure the property of linear phase and causality, we can simply force a symmetry property on $h_d(n)$ and $w(n)$. *How?*
- The frequency response of the actual filter is:

$$H(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(e^{j\theta}) W(e^{j(\omega-\theta)}) d\theta$$

- The window effects:

The sidelobes of $W(e^{j\omega})$ cause the ripples in the passband and stopband of $H(e^{j\omega})$;

The mainlobe of $W(e^{j\omega})$ causes the finite transition between the passband and stopband of $H(e^{j\omega})$.

- Good windows should be such

The mainlobe width of $W(e^{j\omega})$ is small;

The peak sidelobe amplitude of $W(e^{j\omega})$ is small.

- Common windows:

Rectangular: $w(n) = 1 \quad 0 \leq n \leq M$

Bartlett (triangular):

$$w(n) = \begin{cases} 2n/M & 0 \leq n \leq M/2 \\ 2 - 2n/M & M/2 < n \leq M \end{cases}$$

Hanning: $w(n) = 0.5 - 0.5 \cos(2\pi n / M)$

Hamming: $w(n) = 0.54 - 0.46 \cos(2\pi n / M)$

Blackman:

$$w(n) = 0.42 - 0.5 \cos(2\pi n / M) + 0.08 \cos(4\pi n / M)$$

- The spectral comparison of the common windows:

See Figure 7.22 (a)-(e) and Table 7.1

Observe the mainlobe width and peak sidelobe amplitude

- Near-optimal window - Kaiser window:

$$w(n) = \frac{I_0 \left[\beta \left(1 - \left[\frac{(n - \alpha)}{\alpha} \right]^2 \right)^{1/2} \right]}{I_0(\beta)} \quad 0 \leq n \leq M$$

$$\alpha = M / 2$$

$I_0(\cdot)$: zeroth-order modified Bessel function of the 1st kind

M and β control the mainlobe width and peak sidelobe amplitude (see p474)

- Minimax method – optimal filter design:

$$\min_{h(n): 0 \leq n \leq M/2} \left(\max_{\omega \in F} |E(\omega)| \right)$$

$$E(\omega) = W(\omega) \left[H_d(e^{j\omega}) - H(e^{j\omega}) \right]$$

- A key property of an optimal filter:

Equiripple in passband and stopband